Simulation and modeling of BEGe detectors

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MaGe meeting, January 18th 2010





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1 The BEGe detectors

2 The simulation

- The structure of the simulation
- Design and implementation of the simulation

Validation of the simulation

- Validation of the MaGe simulation
- Validation of the PSS

4 Conclusion

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The BEGe geometry



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The (LNGS) BEGe features

		105		digital		104			
Electrical Characteristics:		10 ⁴	-			10 ³ 10 ²			4
Depletion voltage Operational bias voltage	+3000 V +3500 V	51 gn 10 ³		4	\sim	110	0 1173 1	250 1332	
Integral nonlinearity	< 0.05%	10 ²	Ē			Martine 1			
Physical Characteristics:		10 ¹	Ē			Philippy	Many mar should be	i alug 1 - Kagalia	d.
Active diameter Active area Thickness	71 mm 3800 mm ² 32 mm	2.4	500)	1000	1500 Energy [keV]	2000	2500	
Distance from window Efficiency	5 mm > 34%	2.2	0 0	analogue digital	2	I		-0	-
Energy Resolution at 1332.5	keV:	1.8 <u>لح</u> 1.6 ح	_			00			-
FWHM (nominal) FWHM (measured) FWTM	1.752 keV 1.607 ± 0.003 keV 3.259 keV	H 1.4 H 1.2 1 0.8	- - - 69/		0	• •	fitting funct $f(x) = \sqrt{0}$ $f(0) = \sqrt{a}$	ion: 31 + 0.0018 × ~ 0.55 keV	;
			0	500	1000	1500 Energy [keV]	2000	2500	300

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analogue

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1.607 keV_

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I. MC simulation

-> coordinates and energy of the hits

II. Signal formation and development

- $<\!\!-$ coordinate of each hit
- -> electron and hole trajectories
- -> the signal induced on the point size electrode

III. DAQ simulations

- <- energy and signal for each hit in an event
- <- the Preamplifier Transfer Function (PTF)
- -> each pulse is convolved with the PTF
- -> all the pulses of an event are added up
- -> the noise is added to the total pulse

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The simulation design

step 0. Create a library of pulses:

0.1 divide the detector in cubic cell $(1 \text{ mm} \times 1 \text{ mm} \times 1 \text{ mm})$

and generate a pulse for each cell

- 0.2 convolve each pulse with the PTF
- 0.3 store all the pulses in a library
- step 1. Run the MC simulation
- step 2. For each hit compute the pulse as weighted average of the pulses stored in the library
- step 3. For each event compute the total pulse by adding up the pulse of each hit
- step 4. Add the noise

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MaGe: BEGe geometry used munichteststand/GELNGSBEGeDetector.hh

The simulation

MGS: simulation of the signal formation and development

Trajectory simulation

Fourth–order Runge–Kutta method ($\Delta t = 1$ ns):

$$\mathbf{r}(t + \Delta t) = \mathbf{r}(t) + f(\mathbf{v}(\mathbf{r}(t)), \Delta t)$$

where the velocity is computed by using the mobility model of L. Mihailescu and B. Bruynell:

$$\mathbf{v}_h = \mu_h(\mathbf{r}, \mathbf{E}) \cdot \mathbf{E}$$
 $\mathbf{v}_e = \mu_e(\mathbf{r}, \mathbf{E}) \cdot \mathbf{E}$

Simulation of the Electric Field

SOR and relaxation method to solve the Poisson's eq:

$$abla^2 \phi(\mathbf{r}) = -rac{
ho(\mathbf{r})}{arepsilon} \ o \ \mathbf{E}(\mathbf{r}) = -
abla \left(\varphi(\mathbf{r})
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- cathode at 0 V, anode at 3500 V

- detector completely depleted: $\rho(\mathbf{r}) = eN_A(\mathbf{r})$

Signal computation

Shockley-Ramo Theorem:

$$Q(t) = -q\phi_w(\mathbf{r}(t))$$

where $\phi_w(\mathbf{r}(t))$ is the weighting potential





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The weighting potential is defined as the electric potential calculated when the considered electrode is kept at a unit potential, all other electrodes are grounded and all charges inside the device are removed.

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Validation of the MaGe simulation



-> Dead layer measurements (nominal dead layer 0.8 mm) ratio between the counts in the peaks at 81 keV and at 356 keV of $^{133}{\rm Ba}$:



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The validation was carried out by comparing directly the simulated and the experimental signals:

- ²⁴¹Am colimated source \Rightarrow well localized events close to the detector surface;
- \bullet averaging up the experimental and simulated signals \Rightarrow reduction of noise



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Validation of the PSS

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Validation of the simulation Radial scanning $-> ^{241}$ Am source -> 2 mm collimator -> 600 s acquisitions for each position 700



The holes are dragged to the center of the detector and then drift to the $p+\mbox{ contact}$ with a common trajectory

 \Rightarrow pulse shape discrimination parameter A/E^a depends on the final rising part only which is largely independent of the position of interaction inside crystal

 $^aA \rightarrow \,$ max amplitude of the current pulse; $E \rightarrow \,$ total energy of the event

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Validation of the simulation

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Validation of the simulation <u>Circular Scanning</u>

-> ²⁴¹Am source -> 1 mm collimator -> 500 s acquisitions for each position

We study the rise time as a function of the angle.

-> To observe variations we used the rise time between 1% and 90%



Although the experimental data show a behaviour coherent with the simulation, the agreement is only qualitative.

 \Rightarrow the result is remarkable taking into account the problems related to the identification of the time corresponding to the 1% of the maximum amplitude

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Conclusion Conclusion

Results:

- the simulations performed with the nominal geometry is in reasonable quantitative agreement with the experimental data
- the impact of detector parameters (i.e. geometry description, grid step, impurity distribution, bias voltage, etc.) on the signal pulse shape has been studied and the simulation accuracy could be improved.

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Future works:

- investigate the pulse shape discrimination performances of BEGe detectors by using simulations:
 - compare PS discrimination performance of experimental data with the simulation
 - study the impact of the detector parameters on pulse shape discrimination performances and the robustness of A/E method
 - determine the depletion voltage and the best operational voltage
- validate the simulation with a precise inner scanning of the detector
- generate library for the Phase I detectors and study PSA detector

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- validate the simulation with a precise inner scanning of the detector
- generate library for the Phase I detectors and study PSA detector
- -> We are writing a paper containing these results (March-April)

-> The beta version of the simulation software will be soon uploaded to the MaGe repository.

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BEGe detector



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DAQ systems



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HV scanning



region I [2045 V, 3500 V]: excellent performances, detector full depleted, rise time $\sim 0.5\mu$ s, amplitude ~ 0.3 V.

region II [1860 V, 2045 V]: anomalous behaviour, pulses still fast but their amplitudes four times smaller.

region III [100 V, 1860 V]:

detector partially depleted, charge collection not complete, detector capacitance increment, slower rise time $\sim 5\mu s.$



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Characterization measurements - Linearity



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Characterization measurements - Resolution



Energy [keV]	Analogue D	DAQ system	Digital DAQ system			
	peak counts	FWHM [keV]	peak counts	FWHM [keV]		
1173	259899 (510)	1.529 (0.002)	224857 (506)	1.520 (0.002)		
1332	225023 (474)	1.617 (0.002)	200137 (518)	1.607 (0.003)		

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Characterization measurements - Preamplifier



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Characterization measurements - Preamplifier noise



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Characterization measurements - Preamplifier noise



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Validation of the ${\rm MAGE}$ $% {\rm Simulation}$ - Absorption



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Validation of the ${\rm MAGE}\,$ simulation - Barium spectrum



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Validation of the MAGE simulation - DL



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Mobility model

$$\begin{aligned} \mathbf{v}_{exp} &= \frac{\mu_0 \mathbf{E}}{\left(1 + (\mathbf{E}/\mathbf{E}_0)^{\beta}\right)^{1/\beta}} - \mu_n \mathbf{E} \\ \mathbf{v}_d &= \mathscr{A}(|\mathbf{E}|, T) \sum_j \frac{n_j}{n} \frac{\gamma_j \mathbf{E}_0}{\left(\mathbf{E}_0 \gamma_j \mathbf{E}_0\right)^{1/2}} \\ \mathbf{v}_d &\approx \begin{pmatrix} v_r \\ v_{\theta} \\ v_{\phi} \end{pmatrix} = \mathbf{v}_{100}(E) \begin{pmatrix} 1 - \Lambda(k_0) \sin^4(\theta_0) \sin^2(2\phi_0) + \sin^2(2\theta_0) \\ \Omega(k_0) \left[2\sin^3(\theta_0) \cos(\theta_0) \sin^2(2\phi_0) + \sin(4\theta_0)\right] \\ \Omega(k_0) \sin^3(\theta_0) \sin(4\phi_0) \end{pmatrix} \end{aligned}$$

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Backup slides Drift velocity



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How to compute the electric field in a semiconductor detector

Since a semiconductor detector can be considered as an electrostatic system, the electric field can be computed by solving the following Maxwell's equations or, equivalently, by solving the Poisson's equation:

$$\begin{array}{lll} \nabla\cdot \mathbf{E} &=& \displaystyle\frac{\rho}{\varepsilon} & & \\ \nabla\times \mathbf{E} &=& \displaystyle0 &\Rightarrow & \mathbf{E} = -\nabla\phi \end{array} \right\} \quad \nabla\cdot\nabla\phi = -\displaystyle\frac{\rho}{\varepsilon} \quad \Rightarrow \quad \nabla^2\phi = -\displaystyle\frac{\rho}{\varepsilon} \end{array}$$

To solve the Poisson's equation $\nabla^2 \phi = -\rho/\varepsilon$ and find the potential ϕ we need to know:

- the charge density distribution ρ
- the boundary conditions (the value of ϕ on some surfaces):

 $\phi_0|_{S_{cathode}} = V_{cathode}$ and $\phi_0|_{S_{anode}} = V_{anode}$

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The semiconductor detector functioning is based on the properties of a semiconductor junction:

p-type



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n-type

The junction formation:

I. Spontaneus diffusion e 용송 않 h

because of the difference in the concentration of electrons and holes between the two materials. there is an initial diffusion of the holes towards the n-region and a similar diffusion of electrons towards the p-region

II. Recombination

 $e \bullet \longrightarrow e \bullet h$

the diffusing electrons fill up holes in the p-region while the diffusing holes capture electrons in the n-side

III. Thermodinamic equilibrium $\leftarrow \stackrel{\oplus}{\bullet e} \stackrel{E}{\xrightarrow{\ominus}}$

the recombination creates a net charge distribution inside the seminconducor. This creates an electric field gradient across the iunction whitch halts the diffusion process.

The charge distribution dependence of an external electric field





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The charge distribution dependence of the impurity concentrations



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The charge distribution in a real detector

In a real semiconductor the junction is created between an heavily doped semiconductor and a high-purity semiconductor:



In all the potential computation we will assume that:

- the detector is fully depleted p-type detector $\Rightarrow \rho = -eN_d$ n-type detector $\Rightarrow \rho = eN_a$
- the boundary conditions are: the voltage on the electrodes is defined by the HV supply $\Rightarrow \phi|_{cathode} = 0 \text{ V}$ $\Rightarrow \phi|_{anode} = 3000 \text{ V}$

the detector is enclosed in a vacuum chamber

$$\Rightarrow \phi|_{ext} = 0 \text{ V}$$

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The linear superposition principle – the potential

From the linear superposition principle the potential can be separated into two contribution:

$$\phi(\mathbf{r}) = \phi_0(\mathbf{r}) + \phi_\rho(\mathbf{r})$$

where:

- ϕ_0 is the potential calculated considering only the electrode potentials ($\rho({\bf r})=0~~\forall {\bf r}$)
- ϕ_{ρ} is the potential obtained grounding all the electrodes

The linearity of the Maxwell's equation allows for computing the Poisson's equation for each contribution and then add up all the contribution:

$$abla^2 \phi_0(\mathbf{r}) = 0$$
 with: $\phi_0|_{S_{cathode}} = V_{cathode}$ $\phi_0|_{S_{anode}} = V_{anode}$
 $abla^2 \phi_\rho(\mathbf{r}) = -\rho(\mathbf{r})/\varepsilon$ with: $\phi_0|_{S_{cathode}} = 0$ $\phi_0|_{S_{anode}} = 0$

where S_{anode} and $S_{cathode}$ are the boundary surface of the two electrodes.

The linear superposition principle – the field

Similarly, since the electric field is determined by the *linear* relation $\mathbf{E} = -\nabla \phi$, it can be divided into two components:

$$\mathbf{E}(\mathbf{r}) = \mathbf{E}_0(\mathbf{r}) + \mathbf{E}_\rho(\mathbf{r})$$

where:

•
$$\mathbf{E}_0(\mathbf{r}) = -\nabla \phi_0(\mathbf{r})$$

•
$$\mathsf{E}_{
ho}(\mathsf{r}) = -
abla \phi_{
ho}(\mathsf{r})$$

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The Poisson's equation is solved analytically only in the simplest problem, usually it is solved by using numerical methods. In our simulation we use two algorithms which works on a grid:

- The Successive Over Relaxation (SOR) method converges to a solution replacing at each iteration the current approximated solution at a given grid point by a weighted average of its nearest neighbour on the grid
- the relaxation method converges by replacing at each iteration the current approximated solution with its Taylor expansion computed for each point on the grid.

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Comparison



Simulation and modeling of BEGe detectors

Matteo Agostini (MPIK)